

# Cyclic Redundancy Checking

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September 7, 2001

## 1 The Importance of Error Detection

A transmitted bit can be received in error, due to “noise” on the transmission channel. If we are dealing with voice or video data, the occurrence of errors in a small percentage of bits is quite tolerable, but in many other cases it is crucial that all bits be received intact. If for example we are downloading a binary program file, the program may be unexecutable if even one bit is incorrect. If the destination or source address of a packet has one bit wrong, communication is impossible.

There are many methods which have been developed to detect errors, applied at different levels of the seven-layer model. Of course, no method can detect all errors, but a number of methods in use today are amazingly effective. The one we will discuss here is the famous CRC coding form.

## 2 The CRC Error-Detection Method

The Cyclic Redundancy Checking (CRC) appends a few (typically 16 or 32) bits to the end of the bit string for a message and sends out the extended string. The receiver then performs a computation which would yield 0 if no bits of the message had been in error; if the result is not 0, then the receiver knows that there has been an error in one or more bits. (But if the result *is* 0, this does not necessarily mean there was no error.)

CRC cannot detect all possible errors, but the range of errors which it *can* detect is impressively broad. As networks textbook authors Larry Peterson and Bruce Davie have pointed out, it is quite remarkable, for instance, that a mere 32 bits of CRC are sufficient to provide adequate protection against errors in the 12,000-bit messages which Ethernet can send.

### 3 How CRC Works

Let  $M$  be the message we wish to send,  $m$  bits long. Let  $C$  be a divisor string,  $c$  bits long.  $C$  will be fixed (say hardwired into a serial I/O chip), while  $M$  (and  $m$ ) will vary. We must have that:

- $m > c - 1$ .
- $c > 1$ .
- The first and last bits in  $C$  are 1s.

Our CRC field will consist of a string  $R$ ,  $c-1$  bits long. Here is how to generate  $R$ , and send the message:

1. Append  $c-1$  0s to the right end of  $M$ . These are placeholders for where the CRC will go. Call this extended string  $M'$ .
2. Divide  $M'$  by  $C$ , using mod-2 arithmetic. *Note carefully: This is mod-2, not base-2. These quantities are NOT numbers in the usual sense.*

Call the remainder  $R$ . Since we are dividing by a  $c$ -bit quantity,  $R$  will be  $c-1$  bits long.

3. Replace the  $c-1$  appended 0s in  $M'$  by  $R$ . Call this new string  $W$ .
4. Send  $W$  to the receiver.

For example, say  $C = 1011$  and  $M = 1001101$ . Then we divide as follows:

$$\begin{array}{r}
 \phantom{1011} \overline{1010011} \\
 1011 \overline{)1001101000} \\
 \phantom{1011} \underline{1011} \phantom{000} \\
 \phantom{1011} \phantom{1011} \phantom{000} \underline{1010} \phantom{00} \\
 \phantom{1011} \phantom{1011} \phantom{000} \phantom{1010} \phantom{00} \underline{1011} \phantom{0} \\
 \phantom{1011} \phantom{1011} \phantom{000} \phantom{1010} \phantom{00} \phantom{1011} \phantom{0} \underline{1100} \\
 \phantom{1011} \phantom{1011} \phantom{000} \phantom{1010} \phantom{00} \phantom{1011} \phantom{0} \phantom{1100} \phantom{0} \underline{1011} \\
 \phantom{1011} \phantom{1011} \phantom{000} \phantom{1010} \phantom{00} \phantom{1011} \phantom{0} \phantom{1100} \phantom{0} \phantom{1011} \phantom{0} \underline{101}
 \end{array}$$

We first divide 1011 into 1001 (see above), with a quotient of 1. Yes, that's right! As long as the dividend has at least as many digits as the divisor—both have 4 digits here—then the divisor does indeed “go into” the dividend with a quotient of at least 1. If they have the same number of digits,<sup>1</sup> the quotient is 1, which is placed as the first digit in the overall quotient above the division sign. Our remainder is 0010 (remember, our arithmetic here is mod 2, bitwise, i.e. exclusive-OR), which we write as 10.

<sup>1</sup>The number of digits is counted beginning with the first 1 digit, so for example 0011 is considered only a 2-digit quantity, not a 4-digit one.

Next we extend that 10 by bringing down the next digit in the dividend, a 1, forming 101, that is 0101. 1011 goes into 0101 0 times, so we place this 0 as the next digit in the overall quotient above, and bring down the next digit of the dividend, a 0, making 1010. Now 1011 goes into 1010 1 time, and so on.

The final remainder is 101, which replaces the 000 that we had appended to  $M$  to make  $M'$  (step 3), and the transmitter now sends the string 1001101101.

(You should check that one can “multiply” back to get the original result, i.e.  $1011 * 1010011 + 101 = 1001101000$ . Remember, though, that the multiplication and addition are done on a mod-2 basis, i.e. XOR with no carries.)

The receiver then receives a string  $Y$ , which is hopefully the same as  $W$  but might not be, due to line noise. The receiver divides  $Y$  by  $C$ ; if the remainder is nonzero, the receiver decides  $Y$  had one or more bits in error, while if the remainder is zero, the receiver accepts  $Y$  as correct (though it still might not be).

## 4 Why It Works

Note that the replace operation in step 3 above is equivalent to performing

$$W \leftarrow M' + R \tag{1}$$

Moreover, because we are using mod 2 arithmetic, that operation is also equivalent to

$$W \leftarrow M' - R \tag{2}$$

If one divides, say, 63 by 5, then one gets a remainder of 3. That means that if we subtract that remainder of 3 from the dividend 63, the result 60 will be evenly divisible by 5. Similarly, since  $R$  was our remainder from  $M'/C$ ,  $M' - R$  is exactly divisible by  $C$ . So,  $W$  is divisible by  $C$ .

Let  $E$  denote the **error vector**, a symbolic way to describe which bits in  $W$  get corrupted on their way to the receiver. If for instance the  $i$ -th bit of  $E$  is a 1, this means that there was an error in the  $i$ -th bit of  $W$ . Then the receiver will receive  $Y = W + E$ , again keeping in mind that this is mod-2, bitwise addition.

We hope  $E$  consists of all 0s, of course, but it might not. The receiver will divide  $Y$  by  $C$ ; since  $W$  is evenly divisible by  $C$ , then  $Y$  will be too if  $E = 0$ . So, if the receiver finds that  $Y/C$  has a nonzero remainder, the receiver can be sure that there was an error.

In other words, the CRC will detect all error patterns except those for which  $E$  is evenly divisible by  $C$ . Clever people have found good strings  $C$  for which a remarkably small proportion of  $E$ s are evenly divisible by  $C$ . These  $C$ s have been publicized and are in common use.

Note that  $E$  is an  $m+(c-1)$  bit quantity, so it includes the bits corresponding to the  $R$  portion of  $W$ . So, the system here is even capable of detecting some errors in the CRC field itself!

By the way, when we divide  $M'$ , which is  $m+c-1$  bits long, by  $C$ , which is  $c-1$  bits long, getting a quotient  $Q$  and a remainder  $R$ , then  $Q$  will be  $(m+c-1) - c + 1 = m$  bits long.

## 5 What Kinds of Errors Will CRC Detect?

### 5.1 Foundational Material

#### 5.1.1 Notation

In CRC analysis, it helps to represent bit strings as polynomials. We use the bits as coefficients for (fictitious) powers of  $x$ . For instance, our  $C$  above, 1011, would be denoted by

$$1 \cdot x^3 + 0 \cdot x^2 + 1 \cdot x^1 + 1 \cdot x^0 = x^3 + x + 1 \quad (3)$$

The word “denoted” here is key; the polynomial is just notation, and there is no physical variable  $x$ . But this notation turns out to be an extremely powerful tool for analyzing CRC. We will use  $C(x)$ ,  $E(x)$  and so on to refer to the polynomial versions of the bit strings  $C$ ,  $E$  etc.

The question at hand is which types of  $E$  strings are detectable by which  $C$  strings. Remember, we decide what string to use for  $C$ , say when we design a serial I/O chip or Ethernet interface card. So, we want to choose  $C$  in such a way that the overwhelming majority of  $E$  strings are detectable, i.e. yield a nonzero result when divided by  $C$ .

Recall that  $C$  has 1s as its first and last bits,  $c \geq 2$ . That means that the corresponding polynomial,  $C(x)$ , has  $x^{c-1}$  as its leading term, and 1 as its last term. This is illustrated in Equation (3) in the case of  $C = 1011$ ,  $c = 4$ .

#### 5.1.2 The Prime Factorization Theorem

Below are examples of the many properties CRC theorists have proved. Before presenting those proofs, it would be useful to review some basic facts about prime factorization. Let’s abandon the odd arithmetic in CRC for a moment, and consider ordinary integers  $u$ ,  $v$  and  $w$ , where  $w = uv$ .

As you know, each integer has a prime factorization, e.g.  $120 = 2 \times 2 \times 2 \times 3 \times 5$ . Thus, each of  $u$ ,  $v$  and  $w$  has a prime factorization, say  $u = p_{u1}p_{u2}\dots p_{ur}$ ,  $v = p_{v1}p_{v2}\dots p_{vs}$ , and  $w = p_{w1}p_{w2}\dots p_{wt}$ . But since  $w = uv$ , we know that a more detailed representation of the prime factorization of  $w$  is  $p_{u1}\dots p_{ur}p_{v1}\dots p_{vs}$ . Each of the  $p_{ui}$  and  $p_{vj}$  appears somewhere among  $p_{w1}p_{w2}\dots p_{wt}$ . In other words:

#### Prime Factorization Theorem

If  $w = uv$ , then any prime factor of  $w$  must be a prime factor of either  $u$  or  $v$ .

This simple fact will be useful in our proofs below, even though the entities we are working with are now polynomials rather than integers.

Our proofs below will be by contradiction. We are trying to show that  $C(x)$  does not divide  $E(x)$ , but will suppose that it does, and run into a contradiction.

### 5.1.3 Some Example Categories of Detectable Es

Here is a sampling of what can be proved:

- **Any single-bit error will be detected.**

To see this, note that for such an error pattern,  $E$  will consist of all 0s except for a single 1 bit in some position within the string. So,  $E(x)$  will be  $x^k$  for some  $k$ . Note that this is already in prime-factorization form,  $x$  being the prime factor (i.e. it cannot be broken down further into other factors), repeated  $k$  times.

First consider the case in which  $k > 0$ . Suppose  $C(x)$  were to evenly divide  $x^k$ , i.e.  $x^k = C(x)D(x)$  for some polynomial  $D(x)$ . That would imply that  $C(x)$  and  $D(x)$  both factor into powers of  $x$  too, from the Prime Factorization Theorem. Yet  $x$  could not be a factor of  $C(x)$  due to the presence of the constant term 1 in  $C(x)$ . (Recall that  $C(x)$  is required to have the form  $x^{c-1} + \dots + 1$ )

In the case  $k = 0$ , i.e.  $E(x) = 1$ , then the fact that  $C(x)$  has at least one other term than 1 (namely  $x^{c-1}$ ) means  $C(x)$  could not divide  $E(x)$ .

So, this  $E(x)$  is guaranteed not to be evenly divisible by  $C(x)$ , and thus a single-bit error will be caught by the CRC.

- **Any burst error, i.e. a set of consecutive bit errors, of length at most  $c-1$  will be detected.**

For instance, in our worked-out example above in which  $W = 1001101101$ , if  $E$  is, say  $0000111000$ , with consecutive errors in the fifth, sixth and seventh bits of  $W$ , then such a pattern of errors will be detected.

To prove this, let  $w$  denote the length of  $W$ , which is also the length of  $E$ . Say the most significant 1 in  $E$  is in bit position  $j$ , where the least significant position is called position 0 and the most significant position is numbered  $w-1$ . Denote the length of the burst by  $b$ . Then  $E(x)$  will have the form

$$x^j + x^{j-1} + x^{j-2} + \dots + x^{j-b+1} \quad (4)$$

which can be factored as

$$x^{j-b+1}(x^{b-1} + \dots + 1) \quad (5)$$

Suppose  $C(x)$  were to evenly divide this  $E(x)$ , i.e.  $E(x) = C(x)D(x)$  for some polynomial  $D(x)$ . Note again that each  $x$  factor in  $x^{j-b+1}$  is prime, and as in our previous discussion  $x$  is NOT a prime factor of  $C(x)$ . So, all of  $C(x)$ 's prime factors would have to be among the prime factors of  $x^{b-1} + \dots + 1$ . In other words,  $C(x)$  would have to evenly divide  $x^{b-1} + \dots + 1$ . Yet that is impossible, since  $b$  was given

to be at most  $c-1$  and thus  $x^{b-1} + \dots + 1$  has degree less than that of  $C(x)$  (the former has degree  $b-1$ , the latter degree  $c-1$ ).

So we see  $C(x)$  cannot evenly divide  $E(x)$ , and our claim is proved.

This property of catching burst errors is quite important, as it is often the case that line noise will indeed occur in bursts.

- **As long as we choose  $C(x)$  to be a multiple of  $x+1$ , then any  $E$  having an odd number of 1s will be detected.**

First note that  $W(1)$  is equal to the number of 1s in  $W$ , and when evaluated mod-2, we see that  $W(1)$  will be either 1 or 0, depending on whether  $W(x)$  has an odd or even number of 1s. Let's see what this number will be if  $C(x)$  is a multiple of  $x+1$ . In that situation  $W(x)$  will also be a multiple of  $x+1$ , i.e.  $W(x) = (x+1)z(x)$  for some  $z(x)$ . Since  $1+1=0$ , that implies that  $W(1) = (1+1)z(1) = 0$ . In other words, if  $C(x)$  is divisible by  $x+1$ , then  $W(x)$  is guaranteed to have an even number of 1s, so any  $E$  with an odd number of 1s will be detected.

- **If  $C(x)$  is chosen to be a "primitive polynomial," then all 2-bit errors will be detected as long as  $m \leq 2^{c-1} - c$ .**

A polynomial of degree  $d$  is said to be **primitive** if it has the property that it does not divide  $x^k + 1$  for any  $k < 2^d - 1$ .

Suppose our  $C(x)$  is primitive, so that  $C(x)$  does not divide  $x^k + 1$  for any  $k < 2^{c-1} - 1$ . Suppose  $E$  has 2 errors. Then  $E(x) = x^r(1 + x^s)$  for some  $r$  and  $s$ , where  $r, s < m + c - 1$ .

If  $E$  were nondetectable, i.e. if  $E(x)$  were evenly divisible by  $C(x)$ , that would mean that  $C(x)$  would divide  $1 + x^s$  (we discovered earlier that no  $C(x)$  can divide  $x^r$ ). But this couldn't happen as long as  $s < 2^{c-1} - 1$ . Since  $s$  is less than  $m+c-1$ , that means we can assure that this kind of  $E(x)$  will not be divisible by  $C(x)$  by making sure that  $m + c - 1 < 2^{c-1} - 1$ , i.e. by keeping our message length  $m$  smaller than  $2^{c-1} - c$ .

The famous CRC-16 polynomial is  $(x + 1)(x^{15} + x + 1)$ , where the latter factor is primitive.

## 6 Digital Implementation

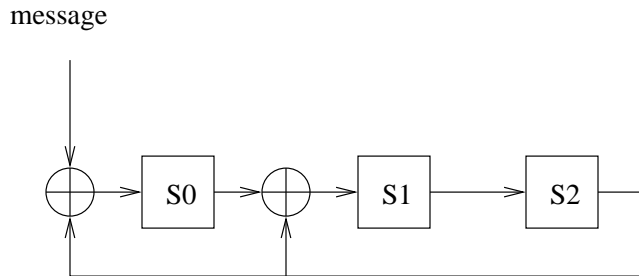
As you can see, the CRC algorithms could easily be encoded in software, but that may be too slow. It would be much faster if implemented in digital logic hardware, and it turns out that this can be done quite simply.

As before, let  $c$  denote the number of bits in the divisor polynomial  $C(x)$ , so that the degree of the polynomial is  $c-1$ . Then we set up a left-to-right shift register which stores  $c-1$  bits, which we will denote, from left to right,  $S_0, S_1, \dots, S_{c-2}$ . As with any shift register, at each clock impulse  $S_i$  is replaced by the value which had been in  $S_{i-1}$ , where we are calling the left input " $S_{-1}$ ."

However, our shift register will differ from an ordinary one in that at each clock pulse the value which had been in  $S_{c-1}$  will be recycled and XOR-ed with some of the other  $S_i$ . Specifically, if there is a term  $x^j$  in the divisor polynomial  $C(x)$ , then there is an XOR just to the left of  $S_i$ .

The message  $M$ , with  $c-1$  0s appended as before, is fed into the input to the shift register, one bit per clock cycle, starting with  $M$ 's most significant (i.e. leftmost) bit. After  $m+c-1$  clock cycles, the shift register contains the remainder (i.e. the CRC field), in reverse order.

Our figure here shows the setup for the case  $C(x) = x^3 + x + 1$ , i.e.  $C = 1011$ , from our example above.



The table here shows a pulse-by-pulse account of what will occur with the input  $M = 1001101$  (with three 0s appended):

clk	input	S0	S1	S2
0	1	0	0	0
1	0	1	0	0
2	0	0	1	0
3	1	0	0	1
4	1	0	1	0
5	0	1	0	1
6	1	1	0	0
7	0	1	1	0
8	0	0	1	1
9	0	1	1	0
end		1	0	1

For example, at the end of clock period 3, the register contains  $(0,0,1)$ , with an input of 1 coming in, say, near the end of the clock period, taking effect in the next period. Then the clock pulse starting period 4 comes. What happens?

- The new  $S_0$  will be the XOR of 1 (from the old  $S_2$ ) and 1 (from the input shown in row 3), which is 0. Thus the new  $S_0$  will be 0.
- The new  $S_1$  will be the XOR of 1 (from the old  $S_2$ ) and 0 (from the old  $S_1$ ), i.e. 1.
- The new  $S_2$  will be equal to the old  $S_1$ , i.e. 0

So, the new register state will be  $(0,1,0)$ , seen in row 4.

The CRC field is calculated to be 101.

This would occur at the transmitter end, and a similar setup for the CRC check will be at the receiver end.